## Chapter 31 Coding Theory

## Motivation

- Use algebraic techniques to protect data transmission affected by noise (human error, imperfect channels, interference etc.)
- Suppose a $(0,1)$ sequence of length of $x_{1} \cdots x_{500}$ is transmitted.
- If there is $1 \%$ probability that $x_{i}$ is transmitted incorrectly for each $x_{i}$, then the probability of correct transmission is $(0.99)^{500} \sim .0066$.
- If each $x_{i}$ is transmitted as $x_{i} x_{i} x_{i}$, and the message is decoded by the maximum likelihood scheme, then the probability for $x_{i}$ to be wrongly decoded is:

$$
3(0.01)^{2}(0.99)^{2}+(0.01)^{3} \sim 0.000298<.0003
$$

- Thus, the probability of correct transmission for each $x_{i}$ is larger than .9997, and the probability of correct transmission of $x_{1} \cdots x_{500}$ is larger than $(.9997)^{500} \sim .86$.
- But repeating many times is not an efficient scheme, so we use algebraic techniques.


## Hamming (7,4) Code

Example Encode $\left(x_{1}, x_{2}, x_{3}, x_{4}\right)$ by $\left(x_{1}, x_{2}, x_{3}, x_{4}\right) G$ with

$$
G=\left[\begin{array}{lllllll}
1 & 0 & 0 & 0 & 1 & 1 & 0 \\
0 & 1 & 0 & 0 & 1 & 0 & 1 \\
0 & 0 & 1 & 0 & 1 & 1 & 1 \\
0 & 0 & 0 & 1 & 0 & 1 & 1
\end{array}\right]
$$

Then we have $2^{4}$ code words in $2^{7}$ such that every pair code words differ in at least 3 digits.

This can be checked by examining the list of nonzero code words in p. 528. Here we need only compute $(x-y) H$ using $\mathbb{Z}_{2}$ arithmetic for any $x-y \neq 0$.

## Linear Code

Definition $\mathrm{An}(n, k)$ linear code over a finite field is a $k$-dimensional subspace $V$ in $\mathbb{F}^{n}$ such that the elements in $V$ are the code words. When $\mathbb{F}$ is $\mathbb{Z}_{2}$, we have a binary code.

Remark The Hamming $(7,4)$ code is a binary code.
Example The set $\{0000,0101,1010,1111\}$ is a $(4,2)$ binary code.
Example The set $\{0000,0121,0212,1022,1110,1201,2011,2102,2220\}$ is a $(4,2)$ linear code over $\mathbb{Z}_{3}$.

## Hamming Distance, Hamming Weight

Definition The Hamming weight $w t(u)$ of $u \in \mathbb{F}^{n}$ is the number of nonzero entries in $u \in \mathbb{F}^{n}$. The Hamming distance $d(u, v)$ of $u, v \in \mathbb{F}^{n}$ is the number of positions in which they differ so that $d(u, v)=w t(u-v)$.
Theorem 31.1 The Hamming distance is a metric (a distance function) in $\mathbb{F}^{n}$.
Proof. (1) $d(u, v) \geq 0$ with equality if and only if $u-v=0$.
(2) $d(u, v)=d(v, u)$.
(3) For $u, v, w$,

$$
d(u, w)=w t(u-w) \leq w t(u-v)+w t(v-w)=d(u, v)+d(v, w) .
$$

To see that $w t(u-w) \leq w t(u-v)+w t(v-w)$, note that if $u_{i}, w_{i}$ are different then $u_{i}, v_{i}$ or $v_{i}, w_{i}$ are different.

Theorem 31.2 Suppose the Hamming weight of a linear code is at least $2 t+1$. Then it can correct any $t$ or fewer errors. Alternatively, it can be used to detect $2 t$ or few errors.

Proof. We use the nearest neighbor decoding.
Suppose $v$ is a received word. Decode it as the nearest code word.
If there is more than one, do not decode. [There are too many errors.]
Suppose $u$ is transmitted and $v$ is received with no more than $t$ errors so that $d(u, v) \leq t$.
Let $w$ be a code word other than $u$.
Then $2 t+1 \leq d(u, w) \leq d(u, v)+d(v, w) \leq t+d(v, w)$
so that $d(v, w)>t$. So, $u$ is the unique correct code word nearest $v$.
Clearly, if $u$ is transmitted as $v$ with fewer than $2 t$ error, then it cannot be another code word.
So, one can detect that are errors in the transmission.

## Standard generator matrix

Remark We cannot use it to do both.
For the Hamming $(7,3)$ code, when there are two errors, one may decode it assuming one error occurs, or assume two errors and refuse to decode.

Systematic code.
We encode a code word $\left(a_{1} \cdots a_{k}\right)$ as $\left(a_{1} \cdots a_{k}\right) G$, where $G=\left[I_{k} \mid A\right]$.
The first $k$ digits are the message digits.
Example Let

$$
G=\left[\begin{array}{llllll}
1 & 0 & 0 & 1 & 1 & 0 \\
0 & 1 & 0 & 1 & 0 & 1 \\
0 & 0 & 1 & 1 & 1 & 1
\end{array}\right]
$$

$$
000000,001111,010101,100110,110011,101001,011010,111100 .
$$

All nonzero code words have weight at least 3 .
So, it will correct single error, or detect up to 2 errors.

Example Example Messages are: $00,01,02,10,11,12,20,21,22$. Let

$$
G=\left[\begin{array}{llll}
1 & 0 & 2 & 1 \\
0 & 1 & 2 & 2
\end{array}\right]
$$

Code words:

$$
0000,0122,0211,1021,1110,1202,2012,2101,2220 .
$$

All code words have weights at least 3 .
So, it will correct single error, or detect up to 2 errors.

Suppose $G=\left[I_{k} \mid A\right]$ is the generator matrix. Let

$$
H=\left[\begin{array}{c}
-A \\
I_{n-k}
\end{array}\right]
$$

be the parity check matrix.

- If $w$ is received, computer $w H$.
- If $w H=0$, then assume no error.
- If $w H$ equals $s$ times the $i$ th row of $H$, then decode $w$ as $w-s e_{i}$, where $e_{i}$ is the $i$ th row of $I_{n}$.
If there are more than one such instance, do not decode. If the code is binary, we simply change the $i$ th position of $w$.
- If the last two cases do not happen, assume more than two errors occur, and do not decode.


## Examples

Use the Hamming $(7,4)$ code with

$$
G=\left[\begin{array}{lllllll}
1 & 0 & 0 & 0 & 1 & 1 & 0 \\
0 & 1 & 0 & 0 & 1 & 0 & 1 \\
0 & 0 & 1 & 0 & 1 & 1 & 1 \\
0 & 0 & 0 & 1 & 0 & 1 & 1
\end{array}\right]
$$

If $v=0000110$ is received, then $v H=110$ which is the first row of $H$.
So, we decode it as 1000110 so that the original message is 1000 .
If $w=1011111$ is received, then $w H=101$, which is the second row of $H$. So, we decode it as 1111111 so that the original message is 1111 .

If $u=1001101$ is sent, and $z=1001011$ is received so that $z H=110$ is the first row. We will wrongly decode it as 0001 .

## Orthogonality relation

Lemma Let $C$ be a systematic $(n, k)$ linear code over $\mathbb{F}$ with a standard generator matrix $G$ and parity matrix $H$. Then $v \in C$ if and only if $v H=0$.

Proof. If $v \in C$, then $v=u G$ so that $v H=u G H=u O=0$.
If $v H=0$, then $v \in \operatorname{Ker}(T)$ where $T: \mathbb{F}^{n} \rightarrow \mathbb{F}^{n-k}$ defined by $x \mapsto x H$.
We will show that the row space of $G$ is $\operatorname{Ker}(T)$ so that $v H=0$ ensures that $v=u G$ for some $u$.

Note that row space of $G$ has dimension $k$, range space of $T$ has $n-k$. So, it suffices to prove that $C \subseteq \operatorname{Ker}(T)$.

It is clear because $G H=0$.

## Parity-Check Matrix Decoding

Theorem 31.3 Parity-check matrix decoding will correct any single error if and only if the rows of the parity-check matrix are nonzero and no one row is a scalar multiple of any other rows.

Proof.

## Coset Decoding

Example Consider the $(6,3)$ binary linear code.

$$
C=\{000000,100110,010101,001001,1110011,101101,011110,111000\}
$$

One can construct the 8 cosets, and use the element with minimum weight as the coset leaders for each of them.
In the above example, the coset leaders (listed as the first column) are: 000000, 100000, 010000, 001000, 000100,000010,000001,100001.

One can decode a received word as the code word in the vertical column containing the received word.

Theorem 31.4 The coset decoding is the same as minimum distance decoding. proof. Let $w$ be a received word.
If $v$ is the coset leader of the coset containing $w$, then $w+C=v+C$.
If $w$ is decoded as $c$, and $c^{\prime}$ is another code word, then

$$
d\left(w, c^{\prime}\right)=w t\left(w-c^{\prime}\right) \geq w t(v)=w t(w-c)=d(w, c)
$$

Thus, $w$ is decoded as $c$, which has a minimum distance to $w$ among all code words.

## Same coset - same syndrome

Definition If an $(n, k)$ linear code over $\mathbb{F}$ has parity-check matrix $H$, then, for any vecotr $u \in \mathbb{F}^{n}$, the vector $u H$ is call the dyndrome of $u$.

Theorem 31.5 Let $C$ be an ( $\mathrm{n}, \mathrm{k}$ ) linear code over $\mathbb{F}$ with a parity-check matrix $H$. Then, two vectors of $\mathbb{F}^{n}$ are in the same coset of $C$ if and only if they have the same syndrome.
Proof. Two vectors $u$ and $v$ are in the same coset of $C$ if and only if $u-v \in C$.
By the orthogonality lemma, $u$ and $v$ are in the same coset if and only if $0=(u-v) H=u H-v H$.

Syndrome decoding for a received word $w$.

1. Compute $w H$, the syndrome.
2. Find the coset leader $v$ such that $w H=v H$.
3. Decode the vector sent was $w-v$.
